

# EEE4001F: Digital Signal Processing

Class Test 1

20 March 2008

## SOLUTIONS

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**Name:**

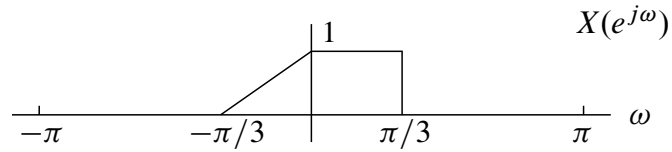
**Student number:**

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### Information

- The test is closed-book.
  - This test has *four* questions, totalling 20 marks.
  - Answer *all* the questions.
  - You have 45 minutes.
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1. (5 marks) A sequence  $x[n]$  has a zero-phase DTFT  $X(e^{j\omega})$  given below:



Sketch the DTFT of the sequence  $y[n] = x[-n]e^{-j\pi n/3}$ .

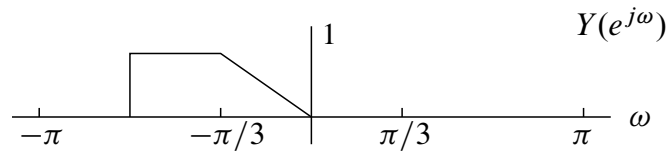
If  $x[n] \iff X(e^{j\omega})$ , then by the time reversal property we know that  $x[-n] \iff X(e^{-j\omega})$ . Applying frequency shift gives

$$x[-n]e^{j\omega_0 n} \iff X(e^{-j(\omega-\omega_0)}),$$

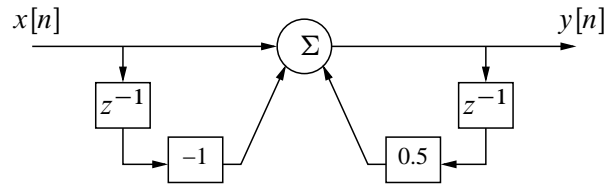
which with  $\omega_0 = -\pi/3$  is

$$x[-n]e^{-j\pi n/3} \iff X(e^{-j(\omega+\pi/3)}).$$

Thus  $Y(e^{j\omega}) = X(e^{-j(\omega+\pi/3)})$ :



2. (5 marks) Consider the following LTI system:



Determine a closed-form expression for the response  $y[n]$  of this system to the following input signal:

$$x[n] = \begin{cases} 1 & n \geq 4 \\ 0 & \text{otherwise} \end{cases}$$

if the system is causal and initially at rest.

The difference equation for the system is

$$y[n] = 0.5y[n-1] + x[n] - x[n-1],$$

so

$$Y(z) = 0.5z^{-1}Y(z) + X(z) - z^{-1}X(z)$$

and

$$H(z) = \frac{Y(z)}{X(z)} = \frac{1 - z^{-1}}{1 - 0.5z^{-1}} = \frac{z - 1}{z - 0.5}.$$

For causality the ROC must be  $|z| > 0.5$ .

The input  $x[n] = u[n-4]$  has z-transform

$$X(z) = \frac{z^{-1}}{1 - z^{-4}}$$

with ROC  $|z| > 1$ , so the output is

$$Y(z) = \frac{1 - z^{-1}}{1 - 0.5z^{-1}} \frac{z^{-4}}{1 - z^{-1}} = \frac{z^{-4}}{1 - 0.5z^{-1}}$$

The ROC of  $Y(z)$  is either  $|z| < 0.5$  or  $|z| > 0.5$ . However, this ROC must contain the intersection of the ROCs of  $X(z)$  and  $H(z)$ , which is  $|z| > 1$ . Thus the ROC for  $Y(z)$  is  $|z| > 0.5$  and the inverse is  $y[n] = (0.5)^{n-4}u[n-4]$ .

3. (5 marks) Consider the following discrete-time signal  $x[n]$ :

$$x[n] = \begin{cases} n + 1 & 0 \leq n \leq 3 \\ 4 & n \geq 4 \\ 0 & \text{otherwise.} \end{cases}$$

- (a) Determine the z-transform  $X(z)$  of  $x[n]$ , and represent it as a ratio of polynomials in  $z^{-1}$ .  
 (b) What is the region of convergence (ROC) of this z-transform?

(a) The z-transform is

$$\begin{aligned} X(z) &= \sum_{n=-\infty}^{\infty} x[n]z^{-n} = 1z^{-0} + 2z^{-1} + 3z^{-2} + 4z^{-3} + \sum_{n=4}^{\infty} 4z^{-n} \\ &= 1 + 2z^{-1} + 3z^{-2} + 4z^{-3} + 4 \sum_{n=4}^{\infty} z^{-n} \end{aligned}$$

Since

$$\sum_{n=4}^{\infty} z^{-n} = z^{-4} + z^{-5} + \dots = z^{-4}(1 + z^{-1} + \dots) = z^{-4} \sum_{n=0}^{\infty} z^{-n},$$

and

$$\sum_{n=0}^{\infty} z^{-n} = \sum_{n=0}^{\infty} (z^{-1})^n = \frac{1}{1 - z^{-1}}$$

for  $|z^{-1}| < 1$ , we have

$$X(z) = 1 + 2z^{-1} + 3z^{-2} + 4z^{-3} + 4 \frac{z^{-4}}{1 - z^{-1}}.$$

This can be written in the form

$$X(z) = \frac{1 + z^{-1} + z^{-2} + z^{-3}}{1 - z^{-1}}.$$

- (b) The previous expression is valid as long as  $|z^{-1}| < 1$ . Letting  $z = re^{j\omega}$  we see that  $|z^{-1}| = |r^{-1}e^{-j\omega}| = r^{-1} = |z|^{-1}$ , so the region of convergence is  $|z|^{-1} < 1$  or  $|z| > 1$ .

4. (5 marks) Consider the continuous-time signal

$$x(t) = \sin(400\pi t + \pi).$$

The discrete-time signal  $x[n]$  is obtained by sampling  $x(t)$  at  $t = n/f_s$  with a sampling frequency  $f_s = 1000$  Hz. Which one of the following continuous-time signals will yield the same sample values when sampled at the same sampling instants? Show and motivate your calculations.

- (a)  $\sin(600\pi t)$
- (b)  $-\sin(1000\pi t)$
- (c)  $\sin(1400\pi t)$
- (d)  $\sin(1600\pi t)$

The discrete time signal obtained by sampling  $x(t)$  is

$$x[n] = x(n/f_s) = \sin\left(\frac{400\pi}{1000}n + \pi\right)$$

which has frequency content at  $\omega = \pm 400\pi/1000 + 2\pi k$ , or  $\omega = \pm 3\pi/5 + 2\pi k$  radians per sample.

Sampling the signal in (a) in the same way gives  $x_a[n] = \sin((600\pi/1000)n)$ , which contains the frequencies  $\omega = \pm 600\pi/1000 + 2\pi k$ , or  $\omega = \pm 3\pi/5 + 2\pi k$  radians per sample. Since  $x[n]$  contains no component at this frequency, the discrete signals cannot be the same. Similarly, the signal in (b) only contains the frequency  $\omega = 0 + 2\pi k$  radians per sample, and the one in (c) only contains  $\omega = \pm 3\pi/5 + 2\pi k$  radians per sample. They also cannot be the same.

Since (d) is the only option, it must be the solution. This can be proven:

$$x[n] = \frac{1}{2j} \left( e^{j[(2\pi/5)n+\pi]} - e^{-j[(2\pi/5)n-\pi]} \right) = \frac{1}{2j} \left( -e^{j(2\pi/5)n} + e^{-j(2\pi/5)n} \right)$$

However,

$$\begin{aligned} x_d[n] &= \sin\left(\frac{1600\pi}{1000}n\right) = \frac{1}{2j} \left( e^{j(8\pi/5)n} - e^{-j(8\pi/5)n} \right) \\ &= \frac{1}{2j} \left( e^{-j(2\pi/5)n} - e^{j(2\pi/5)n} \right) = x[n]. \end{aligned}$$

## Fourier transform properties

Sequences $x[n], y[n]$	Transforms $X(e^{j\omega}), Y(e^{j\omega})$	Property
$ax[n] + by[n]$	$aX(e^{j\omega}) + bY(e^{j\omega})$	Linearity
$x[n - n_d]$	$e^{-j\omega n_d} X(e^{j\omega})$	Time shift
$e^{j\omega_0 n} x[n]$	$X(e^{j(\omega - \omega_0)})$	Frequency shift
$x[-n]$	$X(e^{-j\omega})$	Time reversal
$nx[n]$	$j \frac{dX(e^{j\omega})}{d\omega}$	Frequency diff.
$x[n] * y[n]$	$X(e^{-j\omega})Y(e^{-j\omega})$	Convolution
$x[n]y[n]$	$\frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\theta})Y(e^{j(\omega - \theta)})d\theta$	Modulation

## Common Fourier transform pairs

Sequence	Fourier transform
$\delta[n]$	1
$\delta[n - n_0]$	$e^{-j\omega n_0}$
1 $(-\infty < n < \infty)$	$\sum_{k=-\infty}^{\infty} 2\pi\delta(\omega + 2\pi k)$
$a^n u[n]$ $( a  < 1)$	$\frac{1}{1 - ae^{-j\omega}}$
$u[n]$	$\frac{1}{1 - e^{-j\omega}} + \sum_{k=-\infty}^{\infty} \pi\delta(\omega + 2\pi k)$
$(n + 1)a^n u[n]$ $( a  < 1)$	$\frac{1}{(1 - ae^{-j\omega})^2}$
$\frac{\sin(\omega_c n)}{\pi n}$	$X(e^{j\omega}) = \begin{cases} 1 &  \omega  < \omega_c \\ 0 & \omega_c <  \omega  \leq \pi \end{cases}$
$x[n] = \begin{cases} 1 & 0 \leq n \leq M \\ 0 & \text{otherwise} \end{cases}$	$\frac{\sin[\omega(M+1)/2]}{\sin(\omega/2)} e^{-j\omega M/2}$
$e^{j\omega_0 n}$	$\sum_{k=-\infty}^{\infty} 2\pi\delta(\omega - \omega_0 + 2\pi k)$

## Common z-transform pairs

Sequence	Transform	ROC
$\delta[n]$	1	All $z$
$u[n]$	$\frac{1}{1 - z^{-1}}$	$ z  > 1$
$-u[-n - 1]$	$\frac{1}{1 - z^{-1}}$	$ z  < 1$
$\delta[n - m]$	$z^{-m}$	All $z$ except 0 or $\infty$
$a^n u[n]$	$\frac{1}{1 - az^{-1}}$	$ z  >  a $
$-a^n u[-n - 1]$	$\frac{1}{1 - az^{-1}}$	$ z  <  a $
$na^n u[n]$	$\frac{az^{-1}}{(1 - az^{-1})^2}$	$ z  >  a $
$-na^n u[-n - 1]$	$\frac{az^{-1}}{(1 - az^{-1})^2}$	$ z  <  a $
$\begin{cases} a^n & 0 \leq n \leq N - 1, \\ 0 & \text{otherwise} \end{cases}$	$\frac{1 - a^N z^{-N}}{1 - az^{-1}}$	$ z  > 0$
$\cos(\omega_0 n)u[n]$	$\frac{1 - \cos(\omega_0)z^{-1}}{1 - 2\cos(\omega_0)z^{-1} + z^{-2}}$	$ z  > 1$
$r^n \cos(\omega_0 n)u[n]$	$\frac{1 - r \cos(\omega_0)z^{-1}}{1 - 2r \cos(\omega_0)z^{-1} + r^2 z^{-2}}$	$ z  > r$